## IN THE CLAIMS

1. (Currently amended) An arrangement for iterative channel impulse response estimation in a system employing a transmission channel, comprising:

a channel impulse response estimator (310) for producing <u>iteratively</u> from a received signal ( $^{\hat{y}}$ ) a channel impulse response estimate signal ( $^{\hat{p}}$ ); and

a noise estimator (320) for producing from the received signal (  $^{\mathcal{Y}}$  ) a noise estimate signal,

wherein said noise estimate signal comprises a matrix  $(^W)$  representing the inverse of noise covariance, and

said channel impulse response estimator is arranged, at each iteration (K), to iteratively respond to said matrix (W) to iteratively produce an a single improved channel impulse response estimate signal ( $\hat{P}$ ).

- 2. (Previously presented) The arrangement of claim 1 wherein said matrix (W) representing the inverse of noise covariance is calculated at each iteration.
- 3. (Previously presented) The arrangement of claim 1 wherein said matrix (W) representing the inverse of noise covariance is selected from predetermined values corresponding to statistics of expected noise.
- 4. (Previously presented) The arrangement of claim 2 or 3 wherein the channel impulse response estimate signal  $(\frac{\hat{p}}{2})$  is represented by:

$$(H^H \cdot W \cdot H)^{-1} \cdot H^H \cdot W \cdot \underline{y}$$

where H represents a matrix depending on known symbols,  $\frac{y}{x}$  represents a vector of received channel samples, and W represents the inverse noise covariance matrix.

- 5. (Previously presented) The arrangement of claim 4 when dependent on claim 3 wherein the predetermined values corresponding to statistics of expected noise are selected according to the noise types: Gaussian, upper adjacent interferer, lower adjacent interferer, or co-channel interferer.
- 6. (Currently amended) The arrangement of claim 1 wherein the channel impulse response estimator (310) is arranged to produce the channel impulse response estimate signal  $(\hat{p})$  as a weighted least square function.
- 7. (Previously presented) The arrangement of claim 1 wherein the system is a wireless communication system.
- 8. (Previously presented) The arrangement of claim 7 wherein the system is a GSM system.
- 9. (Previously presented) The arrangement of claim 8 wherein the system is an EDGE system.
- 10. (Previously presented) A receiver for use in a system employing a transmission channel, the receiver comprising the arrangement of claim 1.
- 11. (Currently amended) A method, for iterative channel impulse response estimation in a system employing a transmission channel, comprising:

providing a channel impulse response estimator (310) for producing iteratively from a received signal ( $^{\hat{y}}$ ) a channel impulse response estimate signal ( $^{\hat{p}}$ ); and providing a noise estimator (320) for producing from the received signal ( $^{\hat{y}}$ ) a noise estimate signal,

wherein said noise estimate signal comprises a matrix (W) representing the inverse of noise covariance, and

said channel impulse response estimator, at each iteration (K), (310) iteratively responds to said matrix (W) to iteratively produce an <u>a single</u> improved channel impulse response estimate signal ( $\hat{P}$ ).

- 12. (Previously presented) The method of claim 11 wherein said matrix (W) representing the inverse of noise covariance is calculated at each iteration.
- 13. (Previously presented) The method of claim 11 wherein said matrix (W) representing the inverse of noise covariance is selected from predetermined values corresponding to statistics of expected noise.
- 14. (Previously presented) The method of claim 12 or 13 wherein the channel impulse response estimate signal  $(\hat{p})$  is represented by:

$$(H^H \cdot W \cdot H)^{-1} \cdot H^H \cdot W \cdot \underline{y}$$

where H represents a matrix depending on known symbols,  $\frac{y}{x}$  represents a vector of received channel samples, and W represents the inverse noise covariance matrix.

- 15. (Previously presented) The arrangement of claim 14 when dependent on claim 13 wherein the predetermined values corresponding to statistics of expected noise are selected according to the noise types: Gaussian, upper adjacent interferer, lower adjacent interferer, or co-channel interferer.
- 16. (Currently amended) The method of claim 11 wherein the channel impulse response estimator  $\frac{(310)}{2}$  produces the channel impulse response estimate signal  $\frac{\hat{p}}{2}$  as a weighted least square function.
- 17. (Previously presented) The method of claim 11 wherein the system is a wireless communication system.

- 18. (Previously presented) The method of claim 17 wherein the system is a GSM system.
- 19. (Previously presented) The method of claim 17 wherein the system is an EDGE system.
- 20. (Currently amended) A <u>computer readable medium embodying a</u> computer program element, <u>comprising the</u> computer program <u>means element</u> <u>comprising instructions</u> for performing the <u>a</u> method <u>of claim 11 for iterative channel</u> <u>impulse response estimation in a system employing a transmission channel, the method comprising:</u>

providing a channel impulse response estimator for producing iteratively from a received signal ( $^{\hat{y}}$ ) a channel impulse response estimate signal ( $^{\hat{p}}$ ); and providing a noise estimator for producing from the received signal ( $^{y}$ ) a noise estimate signal.

wherein said noise estimate signal comprises a matrix (W) representing the inverse of noise covariance, and

said channel impulse response estimator, at each iteration (K), responds to said matrix (W) to produce a single improved channel impulse response estimate signal ( $\frac{\hat{p}}{}$ ).